

Correction: A Parallel Primal-Dual Interior-Point Method for DC Optimal Power Flow

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In our PSCC paper, the formulation we used is not equivalent to a DC-OPF problem but rather a network-flow problem. We thank our committee chair Professor Bakirtzis and Professor Chatzivasileiadis for bringing this to our attention. In this note, we update our formulation to a DC-OPF problem and demonstrate the original ideas from our PSCC still hold.

1 DC-OPF Formulation

In order for our method to work well in a distributed setting (*i.e.*, limited communication requirements), it is important that the system for solving the Newton step has a sparse structure. At the same time, the system matrix should be reasonably conditioned so that the iterative solver is well-behaved. For these two reasons, we utilize the following formulation of DC-OPF. Consider a power network with n buses in set \mathcal{N} and m transmission lines in set \mathcal{E} . Let $\theta \in \mathbb{R}^{n-1}$ be the voltage phase angle at each bus except bus 1 (the reference bus). Let b_l be the susceptance of line $l(i, j)$ from bus i to bus j , and \mathbf{b} the vector of all line susceptances.

$$\min_{\theta} \sum_i f_i(p_i) := f(\mathbf{B}\theta) \quad (1)$$

$$\text{s.t. } \underline{\mathbf{p}} \leq \mathbf{B}\theta \leq \bar{\mathbf{p}} \quad (2)$$

$$\underline{\mathbf{f}} \leq \mathbf{D}\mathbf{A}\theta \leq \bar{\mathbf{f}}, \quad (3)$$

where the bus-susceptance $\bar{\mathbf{B}}$ matrix is

$$[\bar{\mathbf{B}}]_{ij} = \begin{cases} -b_l, & \text{if line } l \text{ connects buses } i \text{ and } j \\ 0, & \text{if buses } i \text{ and } j \text{ are not directly connected} \\ -\sum_{j \neq i} [\bar{\mathbf{B}}]_{ij}, & \text{if } i = j. \end{cases} \quad (4)$$

The matrix \mathbf{B} is obtained by removing the first column from $\bar{\mathbf{B}}$. This results from setting the reference angle at bus 1 to 0. The line-bus incidence matrix $\bar{\mathbf{A}}$ to be of dimension $m \times n$ with $[\bar{\mathbf{A}}]_{li} = +1$ and $[\bar{\mathbf{A}}]_{lj} = -1$ if the start bus of line l is i and the end bus of line l is j . The matrix \mathbf{A} is $\bar{\mathbf{A}}$ with the first column removed. The matrix \mathbf{D} is an $(m \times m)$ -matrix whose l th entry is b_l .

In our simulations, we use a quadratic cost function of the form

$$f(\mathbf{B}\theta) = (\mathbf{B}\theta - \mathbf{p}^*)^T \mathbf{W}(\mathbf{B}\theta - \mathbf{p}^*), \quad (5)$$

where \mathbf{p}^* is a vector whose entries are the nominal power demand for demand buses and zero for generation buses. The weighting matrix \mathbf{W} is diagonal and allows to parametrize the different generator costs and demand utility functions.

2 Proposed Method and Iterative Linear Solver Convergence

We verify that our proposed method for calculating the Newton step holds under the DC-OPF formulation in (1)-(3). The Newton step requires solving

$$\begin{bmatrix}
2\mathbf{B}^T\mathbf{W}\mathbf{B} & \mathbf{B}^T & -\mathbf{B}^T & (\mathbf{D}\mathbf{A})^T & -(\mathbf{D}\mathbf{A})^T \\
-\lambda_{\bar{p}}\mathbf{B} & \text{diag}(\mathbf{B}\boldsymbol{\theta} - \bar{\mathbf{p}}) & \mathbf{0} & \mathbf{0} & \mathbf{0} \\
\lambda_{\underline{p}}\mathbf{B} & \mathbf{0} & \text{diag}(\underline{\mathbf{p}} - \mathbf{B}\boldsymbol{\theta}) & \mathbf{0} & \mathbf{0} \\
-\lambda_{\bar{f}}\mathbf{D}\mathbf{A} & \mathbf{0} & \mathbf{0} & \text{diag}(\mathbf{D}\mathbf{A}\boldsymbol{\theta} - \bar{\mathbf{f}}) & \mathbf{0} \\
\lambda_{\underline{f}}\mathbf{D}\mathbf{A} & \mathbf{0} & \mathbf{0} & \mathbf{0} & \text{diag}(\underline{\mathbf{f}} - \mathbf{D}\mathbf{A}\boldsymbol{\theta})
\end{bmatrix}
\begin{bmatrix}
\Delta\boldsymbol{\theta} \\
\Delta\lambda_{\bar{p}} \\
\Delta\lambda_{\underline{p}} \\
\Delta\lambda_{\bar{f}} \\
\Delta\lambda_{\underline{f}}
\end{bmatrix}
= -
\begin{bmatrix}
\mathbf{r}_{dual} \\
\mathbf{r}_{cent,\bar{p}} \\
\mathbf{r}_{cent,\underline{p}} \\
\mathbf{r}_{cent,\bar{f}} \\
\mathbf{r}_{cent,\underline{f}}
\end{bmatrix}, \tag{6}$$

where $\lambda_{\bar{p}}$ are the dual variables associated with the upper bounds on the power injections, $\mathbf{B}\boldsymbol{\theta} - \bar{\mathbf{p}} \leq \mathbf{0}$. Similar definition applies to the other dual variables. The right-hand side quantities are

$$\mathbf{r}_{dual} = 2\mathbf{B}^T\mathbf{W}(\mathbf{B}\boldsymbol{\theta} - \mathbf{p}^*) + \mathbf{B}^T(\lambda_{\bar{p}} - \lambda_{\underline{p}}) + (\mathbf{D}\mathbf{A})^T(\lambda_{\bar{f}} - \lambda_{\underline{f}}) \tag{8}$$

$$\mathbf{r}_{cent} \stackrel{\text{def}}{=} \begin{bmatrix} \mathbf{r}_{cent,\bar{p}} \\ \mathbf{r}_{cent,\underline{p}} \\ \mathbf{r}_{cent,\bar{p}} \\ \mathbf{r}_{cent,\underline{p}} \\ \mathbf{r}_{cent,\bar{f}} \\ \mathbf{r}_{cent,\underline{f}} \end{bmatrix} = \begin{bmatrix} \lambda_{\bar{p}}(\mathbf{B}\boldsymbol{\theta} - \bar{\mathbf{p}}) - \frac{1}{t}\mathbf{1} \\ \lambda_{\underline{p}}(\underline{\mathbf{p}} - \mathbf{B}\boldsymbol{\theta}) - \frac{1}{t}\mathbf{1} \\ \lambda_{\underline{p}}(\underline{\mathbf{p}} - \mathbf{B}\boldsymbol{\theta}) - \frac{1}{t}\mathbf{1} \\ \lambda_{\bar{p}}(\mathbf{D}\mathbf{A}\boldsymbol{\theta} - \bar{\mathbf{f}}) - \frac{1}{t}\mathbf{1} \\ \lambda_{\underline{p}}(\underline{\mathbf{f}} - \mathbf{D}\mathbf{A}\boldsymbol{\theta}) - \frac{1}{t}\mathbf{1} \end{bmatrix} \tag{9}$$

First, we eliminate the dual variables. Since the equality constraints are linear and separable in $\boldsymbol{\theta}$, this is computationally inexpensive and maintains sparsity of the system. Grouping all of the inequality constraints from (2)-(3) into $\mathbf{g}(\boldsymbol{\theta}) \leq \mathbf{0}$, let

$$\mathbf{g}(\boldsymbol{\theta}) \stackrel{\text{def}}{=} \begin{bmatrix} \mathbf{B}\boldsymbol{\theta} - \bar{\mathbf{p}} \\ \underline{\mathbf{p}} - \mathbf{B}\boldsymbol{\theta} \\ \mathbf{D}\mathbf{A}\boldsymbol{\theta} - \bar{\mathbf{f}} \\ \underline{\mathbf{f}} - \mathbf{D}\mathbf{A}\boldsymbol{\theta} \end{bmatrix}, \quad \mathbf{D}\mathbf{g}(\boldsymbol{\theta}) = \begin{bmatrix} \mathbf{B} \\ -\mathbf{B} \\ \mathbf{D}\mathbf{A} \\ -\mathbf{D}\mathbf{A} \end{bmatrix} \tag{10}$$

The reduced system is

$$\mathbf{C}\Delta\boldsymbol{\theta} = \mathbf{w}, \tag{11}$$

where the system matrix and right-hand side are

$$\mathbf{C} = 2\mathbf{B}^T\mathbf{W}\mathbf{B} + \mathbf{D}\mathbf{g}(\boldsymbol{\theta})^T \text{diag}(-\mathbf{g}(\boldsymbol{\theta}))^{-1} \text{diag}(\boldsymbol{\lambda}) \mathbf{D}\mathbf{g}(\boldsymbol{\theta}) \tag{12}$$

$$\mathbf{w} = -\mathbf{r}_{dual} - \mathbf{D}\mathbf{g}(\boldsymbol{\theta})^T \text{diag}(\mathbf{g}(\boldsymbol{\theta}))^{-1} \mathbf{r}_{cent}. \tag{13}$$

Since $\text{diag}(\mathbf{g}(\boldsymbol{\theta}))$ is diagonal, \mathbf{M} exhibits a sparse structure. The original matrix-splitting scheme can be applied to \mathbf{C} in (12). Consider decomposing the matrix \mathbf{C} into the difference of two matrices \mathbf{M} and \mathbf{N} . Provided the spectral radius $\rho(\mathbf{M}^{-1}\mathbf{N})$ is less than one, let the fixed-point of the sequence

$$\Delta\boldsymbol{\theta}^{t+1} = \mathbf{M}^{-1}\mathbf{N}\Delta\boldsymbol{\theta}^t + \mathbf{M}^{-1}\mathbf{w}, \tag{14}$$

be $\Delta\theta^*$, then $\mathbf{C}\Delta\theta^* = \mathbf{w}$. The splitting (*i.e.* choice of \mathbf{M} and \mathbf{N}) should be designed so that i) the matrix-splitting iterates converge and ii) the iterates in (14) are easy to calculate in a distributed way.

Without loss of generality, assume the bus ids and line ids are consecutively assigned across control areas. The matrix \mathbf{C} can be decomposed into the sum of a block-diagonal matrix, \mathbf{D} , and a matrix containing the remaining off-diagonal entries \mathbf{E} . The consecutive ordering of the bus and line indices described above allows for all entries of \mathbf{C} corresponding to buses within the same control area to be contained within a single diagonal block. Specifically, let

$$D_{ij} = \begin{cases} C_{ij} & \text{if buses } i \text{ and } j \\ & \text{belong to the same control area,} \\ 0 & \text{otherwise} \end{cases}, \quad (15)$$

$$E_{ij} = \begin{cases} C_{ij} & \text{if buses } i \text{ and } j \\ & \text{belong to different control areas,} \\ 0 & \text{otherwise} \end{cases}, \quad (16)$$

yielding $\mathbf{C} = \mathbf{D} + \mathbf{E}$. We use the following matrix-splitting design

$$\mathbf{M} = \mathbf{D} + \tau\bar{\mathbf{E}}, \quad \mathbf{N} = \tau\bar{\mathbf{E}} - \mathbf{E}, \quad (17)$$

where τ is a scalar parameter and $\bar{\mathbf{E}}$ is a diagonal matrix whose i th diagonal entry equals

$$\bar{E}_{ii} \stackrel{\text{def}}{=} \sum_{j \neq i} |A_{ij}|. \quad (18)$$

This is a block-Jacobi scheme modified to be diagonally dominant. We have the following proposition which ensures convergence of the matrix-splitting iterates in (14).

Proposition 1. *Using the splitting in (17), for $\tau \geq \frac{1}{2}$, the iterative updates in (14) converge.*

Proof. The sequence $\{\Delta\mathbf{v}^t\}$ in (14) converges to its limit $\Delta\mathbf{v}^*$ as $t \rightarrow \infty$ if and only if the spectral radius of the matrix $\mathbf{M}^{-1}\mathbf{N}$ is strictly less than 1 [1]. Furthermore, if the sequence converges, the limit $\Delta\theta^*$ is the solution of the system, (*i.e.*, $\mathbf{C}\Delta\theta^* = \mathbf{w}$). In order to have the spectral radius $\rho(\mathbf{M}^{-1}\mathbf{N}) < 1$, it is sufficient to have $\mathbf{C} = \mathbf{M} - \mathbf{N} \succ \mathbf{0}$ and $\mathbf{M} + \mathbf{N} \succ \mathbf{0}$ [2].

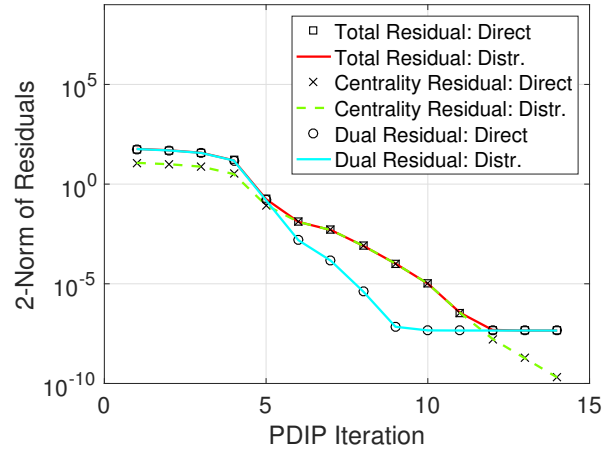
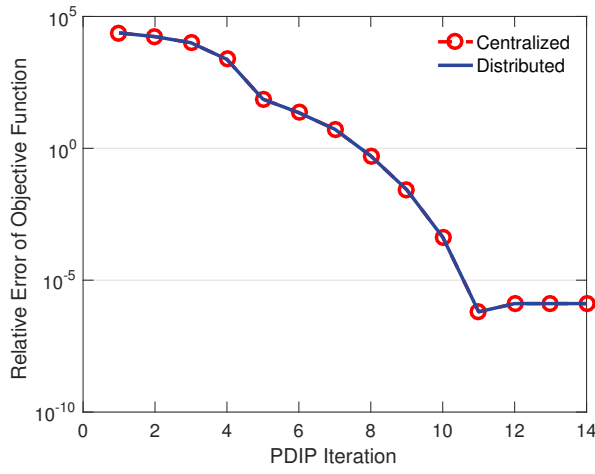
To show that $\mathbf{C} \succ \mathbf{0}$, first we note that the Hessian of the objective function $f_0(\boldsymbol{\theta}) \stackrel{\text{def}}{=} (\mathbf{B}\boldsymbol{\theta} - \mathbf{p}^*)^T \mathbf{W} (\mathbf{B}\boldsymbol{\theta} - \mathbf{p}^*)$ is diagonal and assumed to have strictly positive entries. Then, it is sufficient to show that 1) $D\mathbf{g}(\boldsymbol{\theta})$ has full column rank and 2) $\text{diag}(-\mathbf{g}(\boldsymbol{\theta}))^{-1} \text{diag}(\boldsymbol{\lambda})$ has strictly positive entries. Note from (10) that $D\mathbf{g}(\boldsymbol{\theta})$ has full column rank by construction since \mathbf{B} is the bus susceptance matrix with the column corresponding to the reference angle eliminated. For each constraint i the entries

$$[\text{diag}(\mathbf{g}(\boldsymbol{\theta}))^{-1} \text{diag}(\boldsymbol{\lambda})]_{ii} = \frac{\lambda_i}{-g_i(\boldsymbol{\theta})} > 0 \quad (19)$$

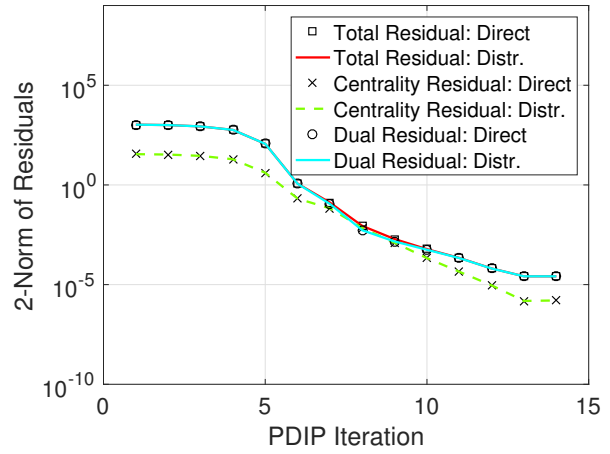
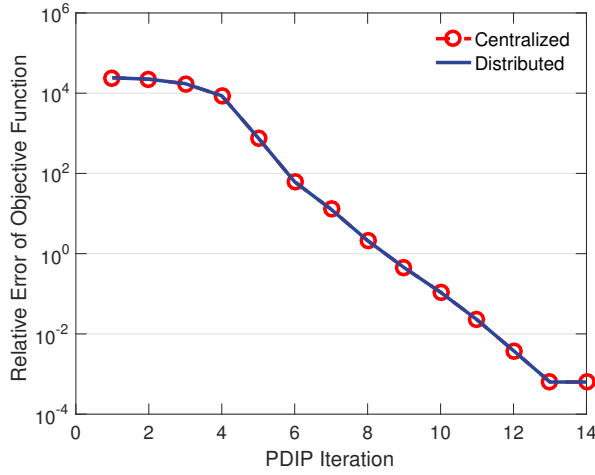
since $\lambda_i > 0$ and $g_i(\boldsymbol{\theta}) < 0$. Lastly, $\mathbf{M} + \mathbf{N}$ is positive definite due to its construction which makes it diagonally dominant. Details on this are provided in [3]. \square

3 Updated Results

In Figure 3 (a)-(d), we show updated results using the new formulation. The IEEE 14-bus system is used in (a) and (b), and the IEEE 118-bus system is used in (c) and (d). The convergence of the relative error of the objective function for the centralized PDIP algorithm (*i.e.*, using direct inversion to calculate the Newton step) is compared to the distributed method (*i.e.*, using the proposed matrix-splitting method to calculate the Newton step). A tolerance of 10^{-10} is set to terminate the matrix-splitting iterates. For 118-bus system, between 10^5 and 10^6 matrix-splitting iterates are required per iteration. This is larger than in the original results using a simpler network-flow problem. While the number of inner-loop iterations required is high, each inner-loop iteration is a computationally inexpensive arithmetic calculation. Future work will be done to study the trade-off between the inner-loop tolerance and the outer-loop convergence. To study the quality of the solution, the convergence of the residuals using our distributed method is shown in (b) and (d) for the 14-bus and 118-bus systems, respectively.



(a) Convergence of Obj. Function Relative Error: 14-Bus System (b) Convergence of Residuals: 14-Bus Study



(c) Convergence of Obj. Function Relative Error: 118-Bus System (d) Convergence of Residuals: 118-Bus Study

References

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